EUROPEAN PATENT APPLICATION

(54) Frequency separation stabilization method for optical heterodyne or optical homodyne communication.

(57) A system for stabilizing a frequency separation between two coherent light sources includes first and second current sources driving first and second semiconductor laser devices (1, 2) with first and second frequencies (f1, f2), respectively. Light from the semiconductor laser devices (1, 2) is combined in a multiplexer (5), and a portion (7) of the combined light is diverted to an optical-to-electrical transducer (8). The output current (10) from the transducer (8) is combined with an output (12) of a local oscillator (11) driven at a selected frequency. A combined signal is then filtered by a filter (14) and used to form a control signal of one of the semiconductor laser devices (2).
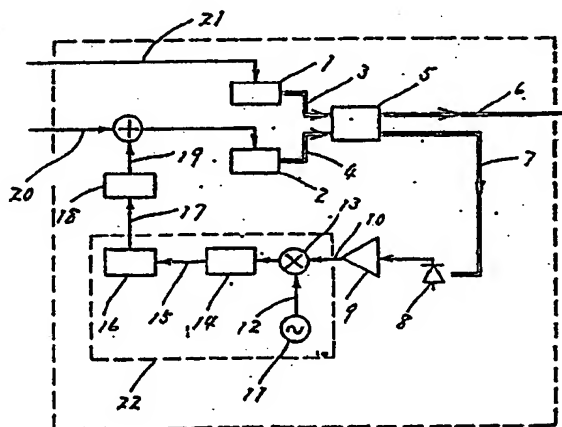
Fig. 1

Fig. 2

EP 0 352 747 A2

FREQUENCY SEPARATION STABILIZATION METHOD FOR OPTICAL HETERODYNE OR OPTICAL HOMODYNE COMMUNICATION

Background of The Invention

This invention relates generally to optical communication systems, and more particularly to systems for stabilizing wavelength or frequency separation of light sources to render them suitable for a wavelength multiplexing or frequency multiplexing systems in coherent optical communication or direct detection optical communication systems.

A conventional wavelength or frequency separation stabilization systems for light sources is discussed in IEEE, International Conference on Communications, Conference Record 37.4.1, pp. 1219 - 1223 (1988). This paper is directed to fiber type Fabry-Perot resonators, or in other words, a kind of etalon, as a means for stabilizing frequency separation. This etalon is used as an optical frequency discriminator for converting an optical frequency to an intensity. The system stabilizes the optical frequency separation by detecting the change of the intensity of the output beam of the etalon that occurs when the optical frequency changes, and by using this signal as a control signal.

In accordance with the technique described above, frequency stability of the light sources is limited by stability of the etalon with respect to the changes in temperature, mechanical vibration, air pressure, and the like. It is therefore difficult to maintain stability for extended periods.

A system is therefore desirable which can facilitate a frequency separation stabilization which is stable for extended periods of time.

Wavelength division multiplexing or frequency division multiplexing optical communication systems using the above-described technique render another problem in the part of a photo-coupler output for combining the beams from a plurality of light sources is discarded without using it. The optical output is not, therefore, utilized effectively.

It is also therefore desirable that a system be provided which can accomplish wavelength division multiplexing or frequency division multiplexing optical communication which utilizes effectively optical outputs of light sources.

Summary of the Invention

In accordance with the present invention, output beams of two light sources are combined and a portion of the combined beam is subjected to optical heterodyne detection by an optical detector and converted to an electric signal. This electric signal and the output of an oscillator are multiplied, and a difference frequency component generated by this multiplication between the electric signal and the output of the oscillator is extracted. The change of this signal component is detected by a frequency discrimination circuit and this signal is applied to one or both of the semiconductor lasers so as to make the difference of the optical frequencies of the two semiconductor lasers constant.

Brief Description of the Drawings

The following will be described in conjunction with the following Figures which form a part hereof and wherein:

Fig. 1 shows A first embodiment of the present invention;

Fig. 2 shows construction of a frequency discrimination circuit of Fig. 1;

Fig. 3 illustrates frequency discrimination characteristics of the frequency discrimination circuit of Fig. 2;

Fig. 4 is a structural view of the light sources of the second and third embodiments of the present invention;

Fig. 5 shows a fourth embodiment of the present invention;

Fig. 6 shows a fifth embodiment of the subject system;

Fig. 7 shows a sixth embodiment of the subject system;

Fig. 8 shows a seventh embodiment of the subject system;

Fig. 9 shows a eighth embodiment of the subject system;

Fig. 10 shows a ninth embodiment of the subject system;

Fig. 11 shows a disposition of the optical frequency of the light sources in a tenth embodiment, on a frequency axis;

Fig. 12 shows structures of tenth, and eleventh and twelfth embodiments;

Fig. 13 shows a disposition of the electric signals on the frequency axis in the eleventh embodiment;

Fig. 14 shows a disposition of the frequency of beams from the light sources on the optical frequency axis in the twelfth embodiment;

Fig. 15 shows a structure of a fourteenth embodiment of the subject system;

Fig. 16 shows another structure of the four-

teenth embodiment;

Fig. 17 shows a fifteenth embodiment of the subject system;

Fig. 18 shows a sixteenth embodiment of the subject system;

Fig. 19 shows a seventeenth embodiment of the subject system;

Fig. 20 shows an eighteenth embodiment of the subject system; and

Fig. 21 shows a nineteenth embodiment of the subject system

Detailed Description of the Preferred and Alternate Embodiments

The principle embodiment of the presence invention will be described with reference to Fig. 1. A semiconductor laser suitably forms light sources 1 and 2.

The output beams 3 and 4 of two semiconductor lasers 1 and 2 are combined by an optical multiplexer 5, such as a photo-coupler, and part of the beam 7 of the output is converted to an electric signal by an optical detector 8. At this time, an electric signal of a difference frequency $f_2 - f_1$ component between the optical frequency f_1 and f_2 (it is hereby assumed that $f_2 > f_1$) of the output beams of the two semiconductor lasers occurs. This signal component and the output 12 of an oscillator 11 are multiplied by a multiplier 13. At this time the frequency of the oscillator output 12 is set to $f_s - f_o$, for example. Here, f_s represents a desired optical frequency separation of the two semiconductor lasers. Then, a signal having a frequency component $f_2 - f_1 - f_s + f_o$ appears at the output of the multiplier 13. The output of a filter 14 permitting the passage of this signal component is applied to a frequency discrimination circuit 6 having a certain specific voltage, e.g. a 0 V output, when the input frequency is f_o. Then, the output of this circuit becomes a signal corresponding to the frequency $f_2 - f_1 - f_s$. It is possible to establish the relation $f_2 - f_1 = f_s$ by applying this signal to the semiconductor laser.

In Fig. 1, the output signal 17 of the frequency discrimination circuit is superimposed with a driving current 20 of the light source 2 to attain this relationship. In this manner, the frequency separation between the two light sources can be kept at the constant value f_s.

In accordance with the present invention, the stability of the frequency separation between the two light sources is determined by the oscillator and the frequency discrimination circuit. This stability is sufficiently higher than stability of optical components, such as etalon, and a stable fre-

quency separation stabilization system can thus be accomplished.

In accordance with the present invention, frequency separation stabilization is conducted by use of part of the output of the optical multiplexer, and an optical wavelength multiplexing or frequency multiplexing communication system is realized by utilizing effectively the optical output, which output has not been used conventionally.

In optical wavelength multiplexing, or frequency multiplexing communication, it is necessary to keep constant the optical frequency separation between the light sources in order to stably secure transmission quality. Considering the case where semiconductor lasers are used for the light sources, the difference $f_2 - f_1$ between the optical frequency f_1 and f_2 (on the premise that $f_2 > f_1$) of the two semiconductor lasers 1 and 2 is kept at an arbitrary frequency f_s. In ordinary wavelength multiplexing or frequency multiplexing communication, the output beams 3 and 4 of the two light sources described above are combined by an optical multiplexer 5, such as a photo-coupler, and part of the output beam is used as a transmission signal. Here, part 7 of the output beam of the optical multiplexer 5 is converted to an electric signal by an optical detector 8 such as a p-i-n photo-diode or APD (Avalanche Photo Diode). This electric signal contains a frequency component $f_2 - f_1$.

The signal is then amplified by a bandpass amplifier 9 which permits the passage of $f_2 - f_1$. In this case, the amplifier 9 may consist of the combination of a low-pass amplifier with a bandpass filter or a high-pass filter. The output signal 10 of this amplifier 9 and the output 12 of an oscillator 11 are multiplied by a multiplier 13. A double balance mixer is an example of this multiplier 13. The output 12 of this oscillator 11 is set to $f_s - f_o$. Here, f_s represents a desired optical frequency separation between the two semiconductor lasers. Then, the frequency of the output of the multiplier 13 becomes $(f_2 - f_1) \pm (f_s - f_o)$. Among them, the low frequency component $(f_2 - f_1) - (f_s - f_o)$ is taken out by the low-pass or bandpass filter 15. This signal 15 is applied to a frequency discrimination circuit 16. This frequency discrimination circuit 16 is set so that when the frequency of the input signal is f_o, its output takes a certain predetermined value, e.g. 0 V, and when the frequency of the input signal is higher than f_o or lower than f_o, a positive or negative signal is outputted, respectively.

This frequency discrimination circuit is suitably comprised of a signal separation circuit (divider), a delay circuit 23 and a double balance mixer 24 as shown in Fig. 2. At this time the delay time of the delay circuit 23 can be expressed substantially as follows:

$$\tau = (n + \tfrac{1}{2})/(2f_o) \qquad (1)$$

where n is an odd number. Then, the input/output characteristics of this frequency discriminator is as shown in Fig. 3<a> and which exhibits desired characteristics. If the output is V and the input frequency is f, $V \alpha cos(2\pi f \tau)$. The output signal (voltage) 17 of this frequency discrimination circuit is applied to a voltage current conversion circuit 18 to obtain a current signal 19. This signal current 19 is added to a driving current 20 of the semiconductor laser 2. In such a construction, the optical frequency f_2 of the semiconductor laser 2 is controlled by the current so as to satisfy the relation $f_2 - f = f_s$. It will be assumed, for example, that the optical frequency f_1 of the semiconductor laser 1 becomes high for some reason (change of temperature, current 20, etc.) and $f_2 - f_1 < f_s$. Then, the frequency $!(f_2 - f_1) - (f_s - f_0)!$ of the input signal 15 to the frequency discrimination circuit 16 becomes smaller than f_0 and the output voltage 17 from the frequency discrimination circuit 16 generates a negative voltage. The voltage-current conversion circuit 18 then generates a current 19 corresponding to the signal 17, which is added to the driving current 20 to raise the optical frequency f_2 of the semiconductor laser 2. (In practice, the current 19 in this case is a negative current and makes the driving current 20 small and f_2 high.) When $f_2 f_1 = f_s$ in this manner, the frequency component $!(f_2 - f_1) - (f_s - f_0)!$ of the signal 15 becomes f_0 and the output voltage 17 of the frequency discrimination circuit becomes 0 V. The optical frequency difference $f_2 - f_1$ between the two semiconductor lasers 1 and 2 is kept at f_s.

The same discussion as described above will now be made about the case of $f_1 > f_2$. In this case, $f_1 - f_2$ must be kept at the predetermined value f_s. The low frequency component 15 of the multiplier 13 becomes $!(f_1 - f_2) - (f_2 - f_0)!$. Therefore, when the signal is added to the driving current 20 of the semiconductor laser 2 in the same way as in the above-mentioned case ($f_2 > f$), the frequency difference $f_1 - f_2$ deviates from f_s. When, for example, f_1 becomes high, $f_1 - f_2 - f_s > 0$ and the output 17 of the frequency discrimination circuit 16 generates a positive voltage so that the optical frequency f_2 of the semiconductor laser 2 becomes low. Accordingly, $f_1 - f_2$ becomes all the more greater and comes away from f_s. In this case, $f_1 - f_2$ can be set to the predetermined value f_s by employing any one of the following four methods.

(1) The signal current 19 is added to the driving current 21 of the semiconductor laser 1, but not to the driving current 20 of the semiconductor laser 2.

(2) The frequency of the output signal 12 of the oscillator 11 is set to $f_s + f_0$.

(3) The polarity of the output of the frequency discrimination circuit is inverted. In other words, a negative signal is outputted when the input frequency is higher than f_0 and a positive signal is outputted when it is lower than f_0. This can be accomplished by setting the delay time τ of the delay circuit 23 in the frequency discrimination circuit 16 of Fig. 2, for example, so as to satisfy substantially the following equation:
$$\tau = (n + \tfrac{1}{2})/(2f_0)   (2)$$
where n is an even number.

The input/output characteristics of the frequency discrimination circuit 16 at this time are such as shown in Fig. 3<b>.

(4) The polarity of the voltage-current conversion circuit 18 is inverted. In other words, a negative current is outputted (current suction) when the input voltage is positive and a positive current is outputted when the input voltage is negative.

Incidentally, this embodiment can be applied similarly to a multi-electrode semiconductor laser. In this case, the signal current 19 may be applied directly to one or a plurality of electrodes of the multi-electrode laser.

Since this embodiment is the system which applies negative feedback to the driving current, the embodiment can accomplish a frequency separation stabilization system which provides a quick frequency response of the feedback system and can follow a sharp optical frequency change.

Next, the second embodiment will be described with reference to Fig. 4. As light sources 1 and 2, this embodiment uses a combination of a semiconductor laser 25, an optical system 26, and a mirror 27 shown in Fig. 4<a>; the combination of the semiconductor laser 25 and an optical fiber 28 shown in Fig. 4<b>; or the combination of the semiconductor laser 25, an optical system 24 and a diffraction grating 29 shown in Fig. 4<c>. At this time the driving current 20' on which the control current 19 is superposed is applied to the semiconductor laser 25. This embodiment provides the effect that a light source having a narrow spectral line width can be obtained.

The third embodiment will be explained with reference to Fig. 4. Though this embodiment uses the combination of the semiconductor laser 25 and the external mirror or the optical fiber or the diffraction grating in the same way as in the second embodiment, this embodiment stabilizes the frequency separation by changing the length 1 between the semiconductor laser 25 and the mirror 27 or the optical fiber 28 or the diffraction grating 29 of the light source so as to change the optical frequency. In the optical source using the diffraction grating, the optical frequency is changed by changing the angle θ as the optical axis by the diffraction grating by use of the output signal so as to stabilize the frequency separation. The distance

1 or the angle θ can be changed by driving the mirror 27, the optical fiber 28, or the diffraction grating 29 by a piezo device. The optical frequency is changed by applying the output signal 17 of the control circuit 22 of Fig. 1 to the driving circuit of the piezo device so that frequency separation can be stabilized. In this embodiment, there can be obtained the light source having a narrow spectral line width.

The fourth embodiment will be explained with reference to Fig. 5. The output voltage 17 of the control circuit 22 is compared with a reference voltage (V_oV in Fig. 5) by a differential amplification circuit 30. Its difference signal 31 is applied to a variable temperature device such as a Peltier device 32 of the light source 2 in order to change the temperature of the light source 2, to change the optical frequency f_2 (generally, the optical frequency drops in semiconductor lasers when the temperature is raised), and to fix the frequency separation $f_2 - f_1$ to the set value f_s. If the output characteristics of the frequency discrimination circuit 15 are set to 0 V when the frequency of its input signal is f_0, the reference voltage Vo of the (+) input terminal of the differential amplification circuit 25 is of course set to 0 V. Incidentally, a laser module 33 incorporating therein integrally the light source 2 and the Peltier device 32 in a casing is hereby used.

Since negative feedback is applied to the Peltier device to change the temperature, this embodiment provides the effect that the breakdown of the device is eliminated due to the drastic change of the driving current of the light source 2 by the control signal. Since the optical frequency has frequency-temperature characteristics of -12 GHz/$^\circ$C in the semiconductor laser, there can be obtained the effect that the optical frequency can be controlled easily by frequency-current characteristics (-1GHz/mA).

The fifth embodiment of the present invention will be explained with reference to Fig. 6. The light sources 1 and 2 are herein assumed to be semiconductor lasers. The optical signal of the light source 1 driven by the driving current 21 passes through the optical fiber as the optical transmission path 3 and is combined with the output beam 4 of the local oscillation light source 2 by the photo-coupler 5. The output 7 of this photo-coupler 5 is converted to the electric signal by the optical detector 8. The frequency component $f_2 - f_1$ (where f_1 and f_2 are optical frequency of the lasers 1 and 2, respectively, and assumed to be $f_2 > f_1$) of this signal is amplified by the amplifier 9 and its part is converted to the baseband signal 35 by the demodulation circuit 34. Part 10 of the output of the amplifier 9 passes through the control circuit 22 and is converted to the control signal 17. As the

control current 19 is applied to the current 20 of the local oscillation light source 2, the optical frequency f_2 of the local oscillation semiconductor laser 2 operates in such a manner as to satisfy the relation $f_2 - f_1 = f_s$. This embodiment stabilizes the intermediate frequency $f_2 - f_1$ in the optical heterodyne communication system. If $f_s = 0$ in this embodiment, the system becomes an optical homodyne communication system.

This embodiment provides the effect that the intermediate frequency can be stabilized and the optical heterodyne communication system and the optical homodyne communication system can be accomplished while keeping stable transmission quality.

In this embodiment, it is possible to use, as the light sources 1 and 2, other lasers; semiconductor lasers having a plurality of electrodes; and a combination of semiconductor lasers, the optical system and the mirror; optical fibers; or diffraction grating shown in Fig. 4. Needless to say, the method of Fig. 5 which changes the temperature of the local oscillation light source 2 by use of the control signal 17 can be applied to this embodiment. When the light sources shown in Fig. 4, are used, the intermediate frequency can of course be stabilized by changing the distance 1 between the semiconductor laser 25 (which serves as the local oscillator light source 2) and the mirror 27 or the optical fiber 28 or the diffraction grating, or changing again the angle θ in the case of the diffraction grating by use of the control signal 17.

The sixth embodiment will be explained with reference to Fig. 7. Fig. 7 is a structural view of an optical heterodyne or optical homodyne system using a balanced type optical receiver. The optical signal 3 transmitted is combined with a local oscillation beam 4 by a 3 dB photo-coupler 5 and after their power is halved, they are outputted as the combined beams 7 and 6. These beams 7 and 6 are incident to optical detectors 8 and 8', respectively, are converted to electric signals and added together to provide one electric signal. Part 10 of this signal is inputted to a control circuit 36 to stabilize the intermediate frequency and the rest are inputted to the demodulation circuit 34 to output a demodulation signal 35.

Since the intensity noise of the local oscillation beam 4 is suppressed, this embodiment provides the effect that a demodulated signal having a high signal-to-noise ratio can be obtained.

The seventh embodiment will be explained with reference to Fig. 8. Fig. 8 is a structural view of an optical heterodyne or optical homodyne system using a polarization diversity type optical receiver 38. The transmitted optical signal 3 is combined with the local oscillation beam 4 by the photo-coupler 5 and its output beam 7 is separated to

beams 7' and 7" having mutually crossing polariza-
tion components by a polarization separator such
as a polarization prism. These beams are con-
verted to electric signals by the optical detectors 8'
and 8, amplified by the amplifiers 9' and 9 and
demodulated by the demodulation circuits 34' and
34 to provide the demodulated signals 35' and 35.
These demodulated signals 35' and 35 are added
to provide a signal 36. The intermediate frequency
is stabilized by the control circuit 36 using part 10
of the output of the amplifier 9. Incidentally, this
embodiment can be accomplished by inputting part
of the output 10' of the amplifier 9' to the control
circuit 36.

This embodiment provides the effect that deg-
radation of the signal-to-noise ratio of the demodu-
lated signal due to the polarization fluctuation of the
optical signal 3 can be suppressed. The degrada-
tion described above can further be suppressed by
furnishing a frequency mixer or an envelope detec-
tor contained in the demodulation circuit 34' or 34
with the amplification square characteristics.

The eighth embodiment will be explained with
reference to Fig. 9. Fig. 9 is a structural view of an
optical heterodyne or optical homodyne system
using a balanced type polarization diversity re-
ceiver 40. The signal beam 3 propagating through
the optical fiber 3 is separated to an x polarization
component beam 3x and a y polarization compo-
nent beam 3y having substantially crossing po-
larization directions with each other by the polariza-
tion separator 39. Examples of the polarization sep-
arator 39 include a TE-TM polarization separator
produced by forming electrodes on an optical
waveguide obtained by diffusing ions of Ti or the
like in a polarization prism or an optical crystal
exhibiting birefringence such as LiNbO3. The local
oscillation beam 4 is separated into two beams 4x
and 4y whose polarization directions are substan-
tially in agreement with 3x and 3y, respectively, by
the polarization separator 39', and these beams 4x
and 46 are combined with the x polarization com-
ponent beams 3x and 3y by couplers 5x and 5y,
respectively.   •

The method of bringing the polarization direc-
tions of the signal beams 3x and 3y into substantial
conformity with the local oscillation beams 3y and
4y, respectively, comprises the steps of setting the
angle and position of the local oscillator optical
source so that the optical power of the output
beam 4x is equal to that of the output beam 4y,
then observing the detection output of the optical
heterodyne (or optical homodyne), that is, the out-
put amplitude of the amplifiers 9x and 9y while
keeping this state (the relative position between the
polarization beam splitter 39' and the local oscilla-
tor optical source 2), and rotating the polarization
beam splitter 39 or the combination of the polariza-

tion beam splitter 39' with the local oscillator op-
tical source 2 so that the output amplitude be-
comes maximal. According to this method the po-
larization directions of the beams 3x and 4x can be
brought into substantial conformity with the po-
larization directions of the beams 4x and 4y, re-
spectively.

Another method of accomplishing the local os-
cillation beams 4x and 4y comprises inserting a
polarization control device into an optical path be-
tween the local oscillator optical source 2 and the
polarization beam splitter 39' and inserting another
polarization control device into the optical paths
from the two outputs of the polarization beam split-
ter 39' to the inputs of the photo-couplers 5x and
5y. In this case, the beam leaving the local oscilla-
tor optical source 2 controls the polarization direc-
tions so that the output power from the polarization
beam splitter 39' is made equal by the polarization
control device. Control is effected by the polariza-
tion control device so that polarization beam com-
ponent 3x of the signal beam and the polarization
beam directions of the two output beams from the
polarization control device 39' are substantially in
agreement with the polarization beam directions of
the x polarization beam component 3x and y po-
larization beam component 3y. This control method
observes the detection output of the optical hetero-
dyne (or the optical homodyne), that is, the output
amplitudes of the amplifiers 9x and 9y, and sets
the polarization beam control device so that the
output amplitudes become maximal. An example of
the polarization beam control device is an optical
fiber whose shape is turned to a circular coil-like
form.

The two outputs 7x and 6x (or 7y and 6y) of
the two optical couplers 5x and 5y are converted to
the electric signals by the two optical detectors 8x
and 8x' (or 8y and 8y') but at this time, optical
heterodyne (or homodyne) detection occurs. The
electric signals are added and amplified by the
amplifier 9x (or 9y) and the baseband signal 35x
(or 35y) is obtained by the demodulation circuit 34x
(or 34y). However, the demodulation circuit 34x is
not necessary in the case of the optical homodyne
detection.

The demodulation circuit may include a delay
detection circuit having a delay device and a multi-
plier, or an envelope detection circuit having a
bandpass filter and an envelope detector.

The outputs 35x and 35y of the demodulation
circuits 34x and 34y are added to obtain a signal
36.

This embodiment provides the effects of sup-
pressing the intensity noise contained in the local
oscillation beam 4 and of preventing deterioration
of the signal-to-noise ratio. Furthermore, this em-
bodiment provides the effect that the degradation

of the signal-to-noise ratio of the demodulation signal due to the polarization beam fluctuation of the optical signal 3 can be checked. This degradation can be further suppressed by furnishing the frequency mixer or the envelope detector contained in the demodulation circuit 34 or 34' with the amplitude square characteristics.

The ninth embodiment will be explained with reference to Fig. 10. Fig. 10 shows the optical frequency separation stabilization system of four light sources 1, 2, 1' and 2'. The beams of the light sources 1 and 2 are combined by the photocoupler 5. The optical frequency separation between the light sources 1 and 2 is stabilized by utilizing part 7 of the output beam by the same means as that of the first embodiment. Similarly, the optical frequency separation between the light sources 1' and 2' is stabilized by utilizing part 7' of the combined beam by the same means as that of the first embodiment. The optical frequency separation between these two sets of optical sources (1 and 2) and (1' and 2') is stabilized by combining part 6, 6' of the photo-couplers 5 and 5' by a photo-coupler 5'' and by changing the optical frequency of the light source 1 by utilizing part 7'' of the output beam by the same means as that of the first embodiment. Incidentally, the pass band of the bandpass type amplifier 9'' used thereby includes the optical frequency separation between the light sources 1 and 1'.

Frequency separation stabilization of the four light sources is accomplished by the means described above.

Though the frequency separation is stabilized in this embodiment by use of the first embodiment but it can also be accomplished naturally by use of the second, third or fourth embodiment or their combination.

Though the frequency separation stabilization of the four light sources is made in this embodiment, it is of course possible to make the frequency separation stabilization of n light sources in general in the same way as described above.

Since the optical frequency separation stabilization of a plurality of optical sources is made in accordance with this embodiment, this embodiment provides the effect that the stable optical transmission apparatus 41 can be accomplished.

Besides the same effect as that of the first embodiment, this embodiment provides also the effect that the optical frequency gap between a plurality of light sources can be stabilized.

The tenth embodiment will be described with reference to Figs. 10 and 11. Means for stabilizing the optical frequency separation between a plurality of (four in Fig. 10) light sources is the same as that of the ninth embodiment but this embodiment is characterized in that the disposition of these light sources is set in such a manner that their optical frequencies on the optical frequency axis are adjacent to one another. In the case of Fig. 10, the optical frequencies of the light sources 1, 2, 1' and 2' are arranged as shown in Fig. 11 <a> or <b>. Symbols f_1 f_2 f_1' and f_2' represent the optical frequencies of the light sources 1, 2, 1' and 2', respectively.

This embodiment provides an effect wherein it is not necessary to broaden the band width of the optical detectors 8, 8' and 8'' and the bandpass filter type amplifiers 9, 9', 9'' shown in Fig. 10.

The eleventh embodiment will be explained with reference to Fig. 12. This embodiment represents means for keeping constant the optical frequency separation between the four light sources 1, 2, 1' and 2'. The beams 3, 4, 3' and 4' from the four light sources are combined by the optical multiplexers 5, 5', 5'' and part 7'' of its output is converted to the electric signal by the optical detector 8. This electric signal is amplified by the bandpass type amplifier 9. Fig. 13 shows the frequency characteristics of the electric signal described above. In this drawing, f_1, f_2, f_1' and f_2' represent the optical frequencies of the light sources 1, 2, 1' and 2', respectively. The band 46 of the amplifier 9 is set so as to amplify only $f_2 - f_1$. The signal processing from here to the acquisition of the signal 19 is the same as that of the first embodiment. This signal passes through the switch 19, and is added to the driving current of the light source 2 so that the optical frequency separation is stabilized between the light sources 1 and 2 by the same operation as that of the first embodiment.

Here, the amplification band of the bandpass amplifier 9 is switched by the control signal 43 from 46 to 46' in Fig. 13. This can be accomplished by changing the passband of the electronically tunable bandpass filter by the control signal 43.

The output signal of this amplifier 9 contains the frequency component $f_2' - f_1'$. This signal is processed in the same way as in the first embodiment. Here, the output signal 12 of the oscillator 11 is switched by the control signal 44 from $f_s - f_0$ to $f_s' - f_0$. Here, f_s' is the target value of stabilization of the frequency separation between the light sources 1' and 2'. The signal 19 is added to the driving current 20' by switching the switch 100 by the control signal 45 so that the optical frequency separation between the light sources 1' and 2' can be stabilized.

Furthermore, the amplification band of the bandpass amplifier 9 is switched by the control signal 43 from 46' to 46'' in Fig. 13. Accordingly, the output signal of the amplifier 9 contains the frequency component $f_1' - f_1$. This signal is processed in the same way as in the first embodi-

ment. Here, the output signal 12 of the oscillator 11 is switched by the control signal 44 from $f_s' - f_o$ to $f_s'' - f_o$. Here, f_s'' is the target value of stabilization of the optical frequency separation between the light sources 1 and 1'. The signal 19 is applied to the driving current 20'' by switching the switch 100 by the control signal 45 so that the optical frequency separation between the light sources 1 and 1' can be stabilized.

The optical frequency separation of all the light sources 1, 1', 2 and 2' can be kept constant by sequentially repeating the simultaneous switching operations of the amplification band of the amplifier 9, the oscillation frequency of the oscillator 11 and the switch 100 by the control signals 43, 44, 45. The control signals 43, 44 and 45 are generated by the controller 42.

This embodiment provides the effect that only one system is necessary as the control system 101 in order to keep the optical frequency gaps between three or more light sources constant.

Needless to say, besides the first embodiment, the second, third and fourth embodiments may be used as the frequency separation stabilization system in this embodiment.

The twelfth embodiment will be explained with reference to Figs. 12 and 14. Means for stabilizing the optical frequency separation between a plurality (four in Fig. 12) of light sources is the same as the one used in the eleventh embodiment but the set values f_s, f_s' and f_s'' of the plurality of optical frequency separation are mutually different as shown in Fig. 14 and this is the characterizing feature of this embodiment. In this drawing, f_1, f_1', f_2 and f_2' are the optical frequencies of the light sources 1, 1', 2 and 2', respectively.

According to this embodiment, since f_s, f_s' and f_s'' are all different from one another, two or three frequencies of $f_1 - f_2$, $f_1' - f_1$ and $f_2' - f_1'$ do not superpose with one another and there can be obtained the effect that the periodic fluctuation of the optical frequencies due to the frequency stabilization signal can be suppressed.

The thirteenth embodiment will be explained. In Fig. 1, when the signal currents corresponding to two independent digital information signals are superposed with the driving currents 20 and 21, respectively, the beams from the semiconductor lasers 2 and 1 come to possess the optical intensity change or phase change corresponding to the digital information signals. Therefore, an optical transmitter 41' for intensity modulation or frequency modulation can be accomplished by causing part 6 of the output of the optical synthesizer 5 to be incident into a transmission optical fiber. This embodiment provides such an optical transmitter.

In this embodiment, the light source may be other lasers, multi-electrode semiconductor lasers,

the combination of the semiconductor laser with the optical system and the mirror, the optical fiber or diffraction grating shown in Fig. 4, and so forth. Stabilization by temperature shown in Fig. 6 may be used as means for stabilizing the frequency separation. Another example of this embodiment is represented by 41' in Fig. 10. Reference numeral 41'' in Fig. 12 represents still another example.

This embodiment provides an optical transmission apparatus for wavelength multiplexing or frequency multiplexing with stabilized frequency separation.

The fourteenth embodiment will be explained with reference to Fig. 15. Reference numeral 41 in this drawing represents the transmission apparatus described in the thirteenth embodiment. The output beam 6 of this optical transmission apparatus 41 passes through the optical fiber 47, and are separated to a beam of each wavelength by the optical multi/demultiplexer 48 and its optical output 50 is received by the optical receiver 49. The optical output 50 of the optical multiplexer/demultiplexer 48 is one of the wavelengths transmitted by transmission apparatus 41 (or optical frequency). The optical multi/demultiplexer produces a number of outputs corresponding to the number of the wavelengths (or frequencies) to be multiplexed which are received from the optical transmission apparatus and in this case, one each optical receiver 49 is necessary for each output. (See Fig. 16 where two receivers are required.)

The optical signal in this embodiment is the intensity-modulated beam and the optical receiver 49 is a direct detection optical receiver consisting at least of the optical detector 51 and an amplifier.

This embodiment provides a wavelength multiplexing or optical frequency multiplexing communication system and a communication apparatus with stabilized optical frequency separation.

Incidentally, other optical transmitters such as 41' in Fig. 1 or 41'' in Fig. 12 may be used as the optical transmitter in Fig. 15.

The fifteenth embodiment will be explained with reference to Fig. 17. The beam 6 from the transmitter 41 of the thirteenth embodiment passes through the optical fiber 47 and the optical output 54 is combined with the beam 55 from the local oscillator optical source 56 by the photo-coupler 53. The output 57 of the photo-coupler 53 is subjected to optical heterodyne or optical homodyne detection by the optical detector 58 and the intermediate frequency signal is amplified by the amplifier 59 and converted to the baseband signal by the demodulation circuit 34. In this manner an arbitrary signal from the frequency-multiplexed beams from the transmitter 41 can be received by the optical heterodyne or optical homodyne receiver 60. Incidentally, the optical signal of this

embodiment is the amplitude-modulated. frequency-modulated or phase-modulated beam.

This embodiment provides a wavelength multiplexing or optical frequency multiplexing communication system and communication apparatus with stabilized optical frequency separation.

Incidentally, other optical transmitters such as 41′ in Fig. 1 or 41″ in Fig. 12 may be used as the optical transmitter in Fig. 17.

The sixteenth embodiment will be explained with reference to Fig. 18. This embodiment is characterized in that a balanced type receiver 61 is used in place of the optical heterodyne or optical homodyne receiver 60 of the fifteenth embodiment. This embodiment relates to the combined system of the transmitter 41 and the receiver 60 and its transmitter. The structure and operation principle of the balanced type receiver are explained in the sixth embodiment.

Incidentally, other optical transmitters such as 41′ in Fig. 1 or 41″ in Fig. 12 may be used as the optical transmitter of this embodiment.

Since the intensity noise of the local oscillation beam 55 is suppressed, this embodiment provides a system capable of obtaining a demodulated signal having a high signal-to-noise ratio, and a transmitter for the system.

The seventeenth embodiment will be explained with reference to Fig. 19. This embodiment uses a polarization diversity receiver 62 as the optical heterodyne or homodyne receiver 62. The structure and operation principle of the polarization diversity receiver are explained already with reference to the seventh embodiment. This embodiment relates to the combined system of the transmitter 41 with the receiver 61 and its transmission apparatus. Incidentally, other optical transmitters such as 41′ in Fig. 1 or 41″ in Fig. 12 may be used as the optical transmitter of this embodiment.

This embodiment effectively prevents degradation of the signal-to-noise ratio of the demodulated signal 36 due to polarization fluctuation of the optical fiber output 54. Furthermore, the embodiment provides a communication system and its promoter capable of further suppressing the degradation of signal to noise can be ratio by furnishing the frequency mixers or the envelope detectors contained in the demodulation circuits 34 and 34′ with the amplitude square characteristics.

The eighteenth embodiment will be explained with reference to Fig. 20. This embodiment uses a balanced type polarization diversity receiver 63 as the optical heterodyne or homodyne receiver. The structure and operation principle of the balanced type polarization diversity receiver are explained already in the eighth embodiment. This embodiment relates to the combined system of the optical transmitter 41 and the receiver 63 and its transmis-

sion apparatus.

Incidentally, the optical transmitter of this embodiment may be other optical transmitters such as 41′ in Fig. 1 or 41″ in Fig. 12.

This embodiment provides the effect that the intensity noise contained in the beam from the local oscillator optical source 56 is extinguished and degradation of the signal-to-noise ratio can be prevented. Furthermore, the embodiment provides the effect that degradation of the signal-to-noise ratio of the demodulated signal due to the polarization fluctuation of the optical signal 54 can be suppressed. Still further, the embodiment provides the effect that the degradation can further be suppressed by furnishing the frequency mixer or the envelope detectors contained in the demodulation circuit 34 or 34′ with the amplitude square characteristics.

The nineteenth embodiment will be explained with reference to Fig. 21. In Fig. 21, n optical fibers 471, 472, ... , 47n are shown extended from a center station 64 to n stations 651, 652, ... , 65n. The center station 64 stores n optical transmitters 411, 412, ... , 41n and these optical transmitters are the aforementioned transmitters 41, 41′ or 41″. At least one optical receiver 601, 602, ... , 60n exists in other station 651, 652, ... , 65n and these optical receivers are the aforementioned optical receiver 49, 60, 61, 62 or 63.

In this embodiment, the wavelength-multiplexed or frequency-multiplexed beams are transmitted from the center station 64 to the other stations 651, 652, ... , 65n and the other stations 651, 652, ... , 65n receive the information by any of direct detection, optical heterodyne detection and optical homodyne detection. This embodiment relates to this system and an apparatus for practicing the system.

Needless to say, subscribers may substitute the other stations 651, 652, ... , 65n in this embodiment. A two-way communication system can of course be accomplished by disposing the optical transmitter in at least one of the other stations 651, 652, ... , 65n so as to transmit the optical signal from the other station to the center station 64 and to receive it by the optical receiver installed in the center station.

This embodiment provides the effect that a one-way or two-way optical communication network having stable optical frequency separation and high transmission quality can be configured.

In accordance with the present invention, the optical frequency separation can be stabilized without using the optical etalon or the like so that an optical frequency stabilization system which is stable for a long period can be accomplished.

Since optical frequency stabilization is made by utilizing part of the photo-coupler in accordance

with the present invention, the beam from the light source can be utilized effectively. In practice, only 1/2 of the photo-coupler output has been utilized conventionally in the wavelength multiplexing or optical frequency multiplexing system, but this embodiment can utilize all the photo-coupler output.

Claims

1. A frequency separation stabilization method comprising the steps of:
multiplexing output beams from two light sources to form a combined beam;
converting at least a portion of the combined beam to an electric signal by an optical detector;
multiplying at least a portion of the electric signal, having a difference component in optical frequency between said light sources, by the output from an oscillator to form a multiplication signal;
extracting a difference frequency component between said electric signal and the output of said oscillator;
converting the difference frequency component to an electric signal by a frequency discrimination circuit; and
changing optical frequencies of said light sources by applying said output signal from said frequency discrimination circuit to one or both of said light sources to keep the optical frequency separation between said two light sources constant.

2. The frequency separation stabilization method according to claim 1, wherein said step of multiplexing light sources includes the step of multiplexing at least one of semiconductor lasers and multi-electrode semiconductor lasers.

3. The frequency separation stabilization method according to claim 1, wherein said step of multiplexing light sources includes the step of multiplexing light sources comprised of semiconductor lasers in conjunction with of at least one of reflecting mirrors and diffraction gratings.

4. The frequency separation stabilization method of claim 3 including the step of changing at least one of an angle of said reflecting mirror and said diffraction grating, in accordance with an output signal of the frequency discrimination circuity, in order to change the optical frequency of said light sources and to keep the optical frequency separation between said two light sources constant.

5. The frequency separation stabilization method of claim 1 further comprising the step of using the output signal of said frequency discrimination circuit as a control signal for changing temperature of at least one of said light sources in order to change an optical frequency of said light sources and to keep the optical frequency separation between said two light sources constant.

6. The method of claim 1 wherein the step of multiplexing output beams includes the step of encoding data for communication, in accordance with one of optical homodyning and optical heterodyning, by controlling a first of the two light sources which functions as a signal light source, and wherein the step of changing the optical frequencies includes the step of keeping an intermediate frequency of multiplexed output beams constant by controlling a second of the two light sources which functions as a local oscillator light source.

7. The method of claim 6, further comprising the steps of:
combining, with a balanced optical receiver having optical couplers, the signal from said signal light source with the beam from said local oscillator light source;
converting, with two optical detectors the two output beams to electric signals; and
adding said two electric signals to each other.

8. The method of claim 6 further comprising the steps of:
combining, with a polarization diversity receiver consisting of an optical coupler, the signal beam with the local oscillation beam and a polarization beam into an output beam;
splitting, with a splitter, the output beam into two substantially crossing polarization components;
converting, with two optical detectors, the two polarization components to two electric polarization signals;
demodulating, with two demodulation circuits, said two electric polarization signals; and
adding, with an adder, said two demodulation signals.

9. The method of claim 6 further comprising the steps of:
separating, with a balanced type polarization diversity receiver, said signal beam from said signal light source into an x polarization component beam and a y polarization component beam crossing each other substantially at right angles;
separating the beam of said local oscillator optical source into two beams substantially in conformity with the polarization directions of said x and y polarization component beams, respectively;
combining, with two optical multiplexers, the two beams substantially in conformity with the polarization directions;
converting, with two optical detectors, two output beams from said optical multiplexer with said x polarization component beam with the beam from said local oscillator optical source into first electric signals;
adding the first electric signals to form a first added signal;
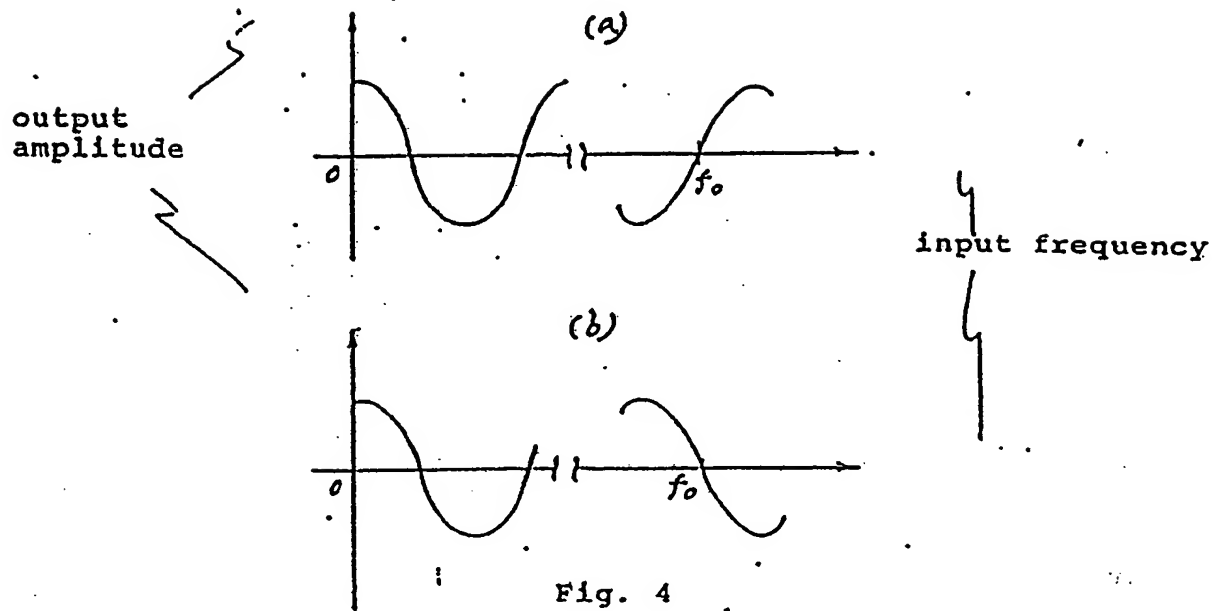demodulating, with a demodulation circuit, said first

Fig. 3

(a)

output
amplitude

f_0

input frequency

(b)

f_0

Fig. 4

(a)

27 26 25

l

28

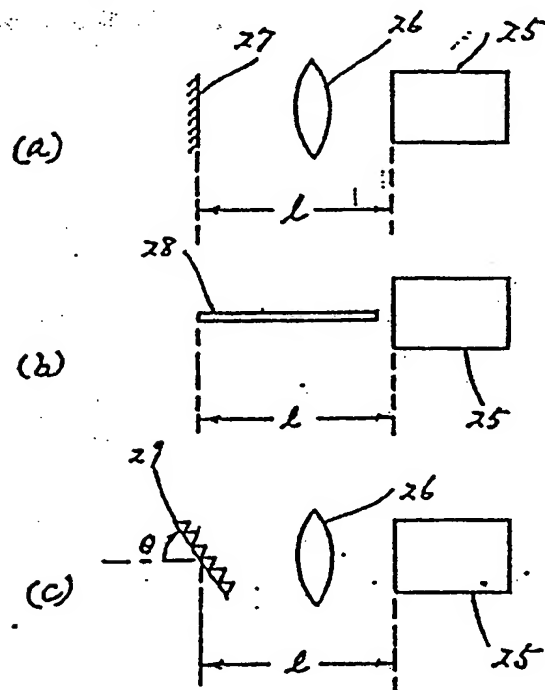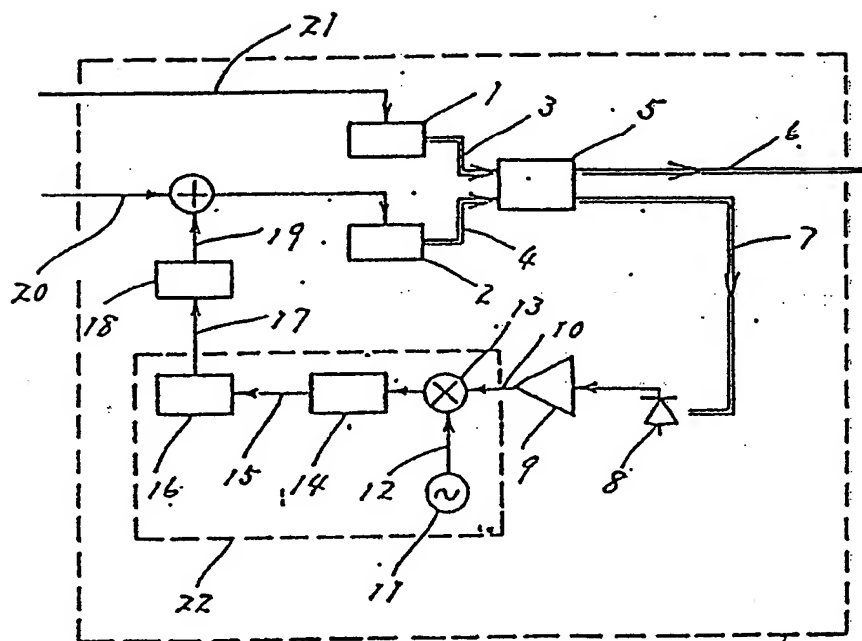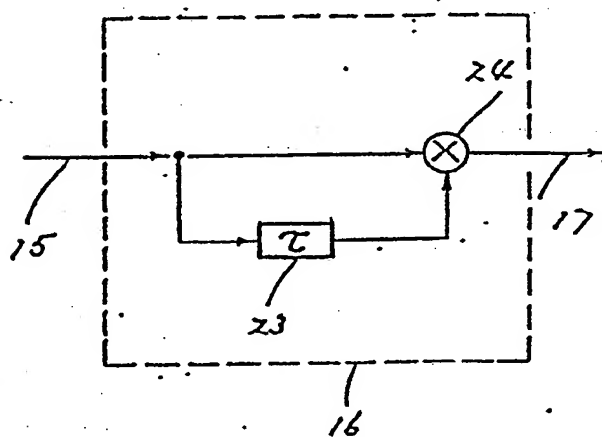(b)

l

25

29

θ

(c)

26

25

l

25

Fig. 1

Fig. 2

added signal to form a first demodulated signal;

converting, with two optical detectors, the two output beams of said optical multiplexer for combining said y polarization component beam with the beam from said local oscillator optical source into second electric signals;

adding said second electric signals to form a second added signal;

demodulating, with a demodulation circuit, said second added signal to form a second demodulated signal; and

adding, with an addition circuit, said first and second demodulated signals to form an added demodulated signal.

10. A system for stabilizing a difference frequency between two coherent light sources comprising:

means for generating a first periodic electric signal having a frequency f1;

means for driving a first light source with the first periodic electric signal such that the first light source has a coherent optical wave output with a frequency f1;

means for generating a second periodic electric signal having a frequency f2;

means for driving a second coherent light source with the second periodic signal such that the second light source has a coherent optical wave output with a frequency f2;

multiplexer means for combining the first and second coherent optical waves into a combined optical wave output having a frequency f2-f1;

means for diverting at portion of the combined optical wave to an optical-to-electrical transducer;

optical-to-electrical transducer means for forming an combined electrical signal from the combined optical wave such that the combined electrical signal has a frequency f2-f1;

oscillator means for generating an electrical oscillator signal having a selected difference frequency fs-fo, wherein fs is a desired frequency separation between f2 and f1, and wherein fo is a desired output frequency;

multiplier means for multiplying the electrical oscillator signal and the combined electrical signal to form a multiplied electrical signal having a frequency (f2-f1) +/- (fs-fo);

means for filtering out a portion of the multiplied signal to form an electrical input signal having a frequency (f2-f1) - (fs-fo);

frequency discrimination means for generating a control signal in accordance with the electrical input signal; and

frequency control means for altering a frequency of at least one of the first and second period electrical signals in accordance with the control signal.

11. The stabilizing system of claim 10 wherein the frequency discrimination means includes a

means for generating a first voltage level when a frequency of the electrical input signal exceeds fo, means for generating a second voltage level when a frequency of the electrical input signal equals fo, and means for generating a third electrical input signal when a frequency of the electrical input signal is less than fo.

12. The stabilizing system of claim 11 wherein the frequency discrimination means is comprised of:

a signal separation circuit means for dividing the electrical input signal into first and second portions;

a delay circuit means for forming a delayed second portion from the second portion of the electrical input signal;

double balance mixer means for forming the control signal from the first portion of the electrical input signal and the delayed portion of the electrical input signal.

13. The stabilizing system of claim 12 wherein at least one of the first and second light sources is comprised of a combination of a semiconductor laser device and a mirror separated by a lens system, and wherein the frequency control means is comprised of means for varying a distance between a semiconductor laser device and its associated mirror in accordance with the control signal.

14. The stabilizing system of claim 12 wherein at least one of the first and second light sources is comprised of a semiconductor laser device and an optical fiber, and wherein the frequency control means is comprised of a means for varying an effective length of the optical fiber in accordance with the control signal.

15. The stabilizing system of claim 12 wherein at least one of the first and second light sources is comprised of a semiconductor laser device and a diffraction grating separated by a lens system, and wherein the frequency control means is comprised of a means for varying at least one of (a) a distance between a semiconductor laser device and its associated diffraction grating, and (b) an angle between a semiconductor laser device and its associated diffraction grating, in accordance with the control signal.

16. The stabilizing device of claim 12 wherein at least one of the first and second light sources is comprised of a semiconductor laser device, and wherein the frequency control means is comprised of a means for varying a temperature of at least one of the semiconductor laser devices in accordance with the control signal.

17. A method for stabilizing a difference frequency between two periodic light sources comprising the steps of:

generating a first periodic electric signal having a frequency f1;

driving a first light source with the first periodic

electric signal such that the first light source has a periodic optical wave output with a frequency f1;

generating a second periodic electric signal having a frequency f2;

driving a second periodic light source with the second periodic signal such that the second light source has a periodic optical wave output with a frequency f2;

combining the first and second periodic optical waves into combined optical wave output having a frequency f2-f1;

diverting at portion of the combined optical wave to an optical-to-electrical transducer;

forming an combined electrical signal from the combined optical wave such that the combined electrical signal has a frequency f2-f1;

generating an electrical oscillator signal having a selected difference frequency fs-fo, wherein fs is a desired frequency separation between f2 and f1, and wherein fo is a desired output frequency;

multiplying the electrical oscillator signal and the combined electrical signal to form a multiplied electrical signal having a frequency (f2-f1) +/- (fs-fo);

filtering out a portion of the multiplied signal to form an electrical input signal having a frequency (f2-f1) - (fs-fo);

generating a control signal in accordance with the electrical input signal;

altering a frequency of at least one of the first and second period electrical signals in accordance with the control signal.

18. An optical communication system comprising:

means for generating a first periodic electric signal having a frequency f1;

means for driving a first light source with the first periodic electric signal such that the first light source has a periodic optical wave output with a frequency f1;

means for generating a second periodic electric signal having a frequency f2;

means for driving a local oscillation light source with the second periodic signal such that the local oscillation light source has a periodic optical wave output with a frequency f2;

photo-coupler means for combining the first and second periodic optical waves into combined optical wave output having a frequency f2-f1;

means for diverting at portion of the combined optical wave to an optical-to-electrical transducer;

optical-to-electrical transducer means for forming an combined electrical signal from the combined optical wave such that the combined electrical signal has a frequency f2-f1;

demodulation means for converting a portion of the combined electrical signal into a baseband signal;

oscillator means for generating an electrical oscilla-

tor signal having a selected difference frequency fs-fo, wherein fs is a desired frequency separation between f2 and f1, and wherein fo is a desired output frequency;

multiplier means for multiplying the electrical oscillator signal and the combined electrical signal to form a multiplied electrical signal having a frequency (f2-f1) +/- (fs-fo);

means for filtering out a portion of the multiplied signal to form an electrical input signal having a frequency (f2-f1) - (fs-fo);

frequency discrimination means for generating a control signal in accordance with the electrical input signal; and

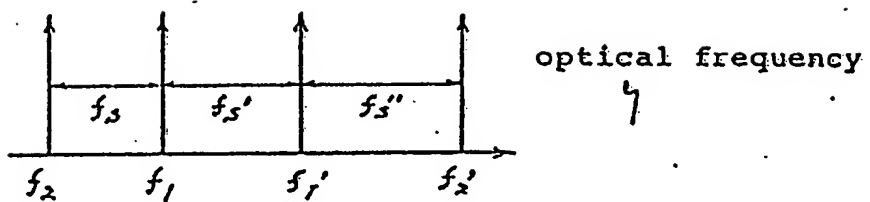frequency control means for altering a frequency of the local oscillation light source in accordance with the control signal.

Fig. 5

Fig. 6

Fig. 7

Fig. 8

Fig. 9

Fig. 10

Fig. 11

⟨a⟩

f_2 f_1 f_1' f_2'

⟨b⟩

f_2' f_1' f_1 f_2

optical frequency

Fig. 13

46   46''   46'

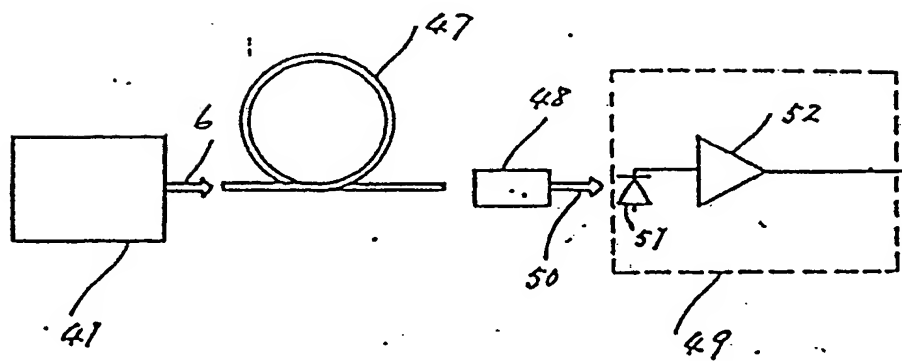$f_1 - f_2$ $f_1' - f_1$ $f_2' - f_1'$
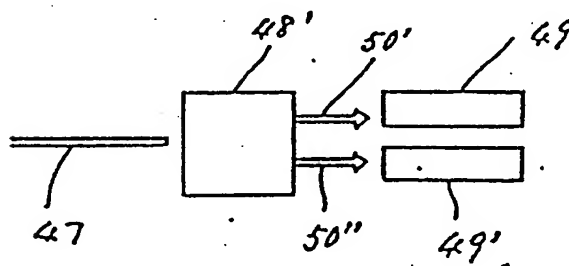
frequency

Fig. 12

Fig. 14

Fig. 15

optical frequency
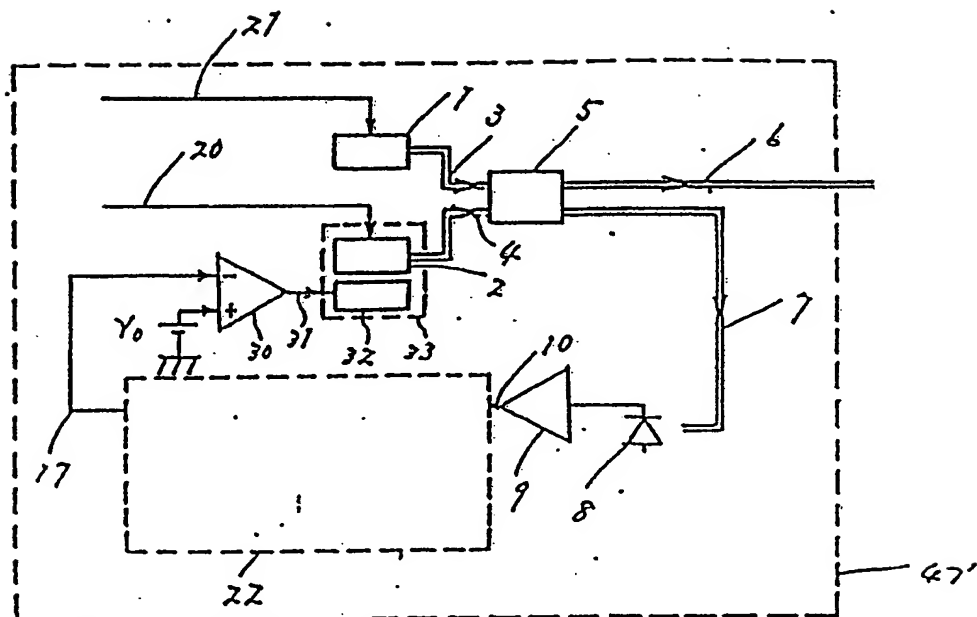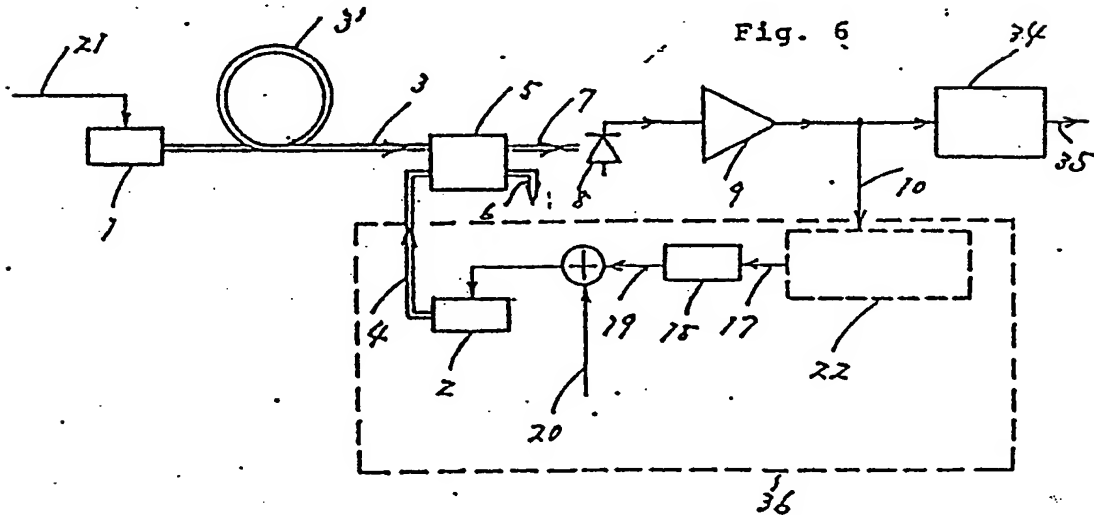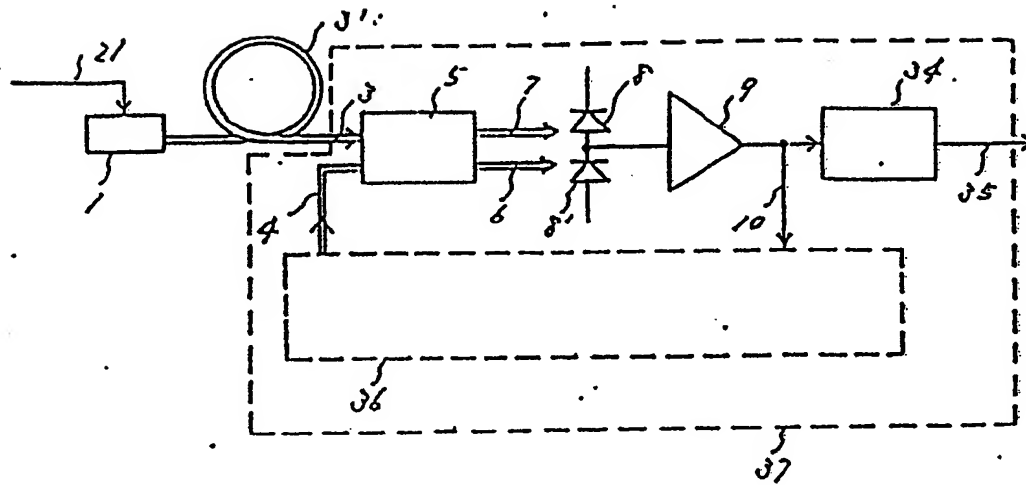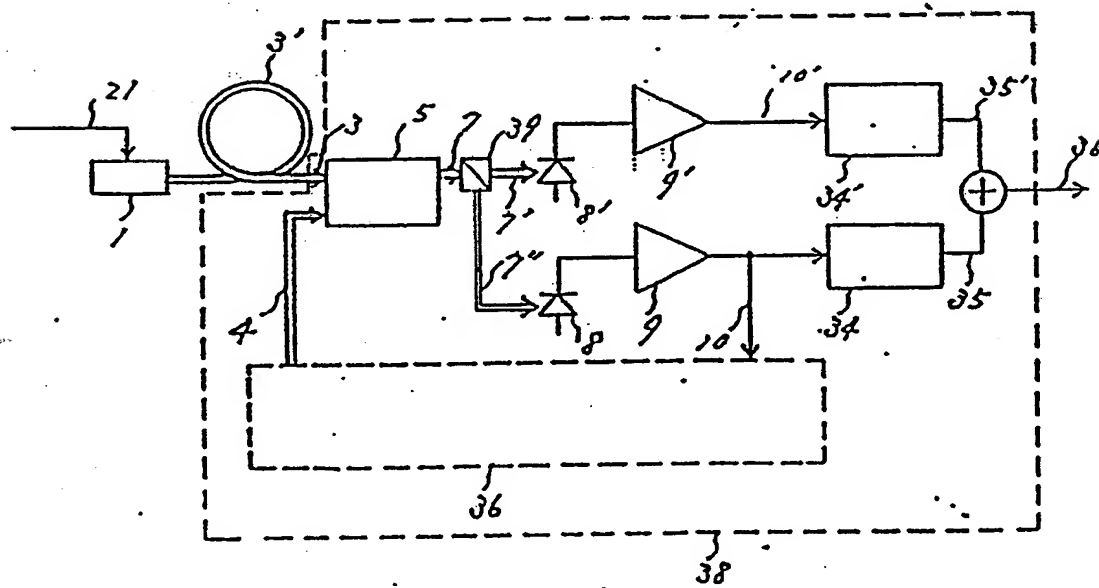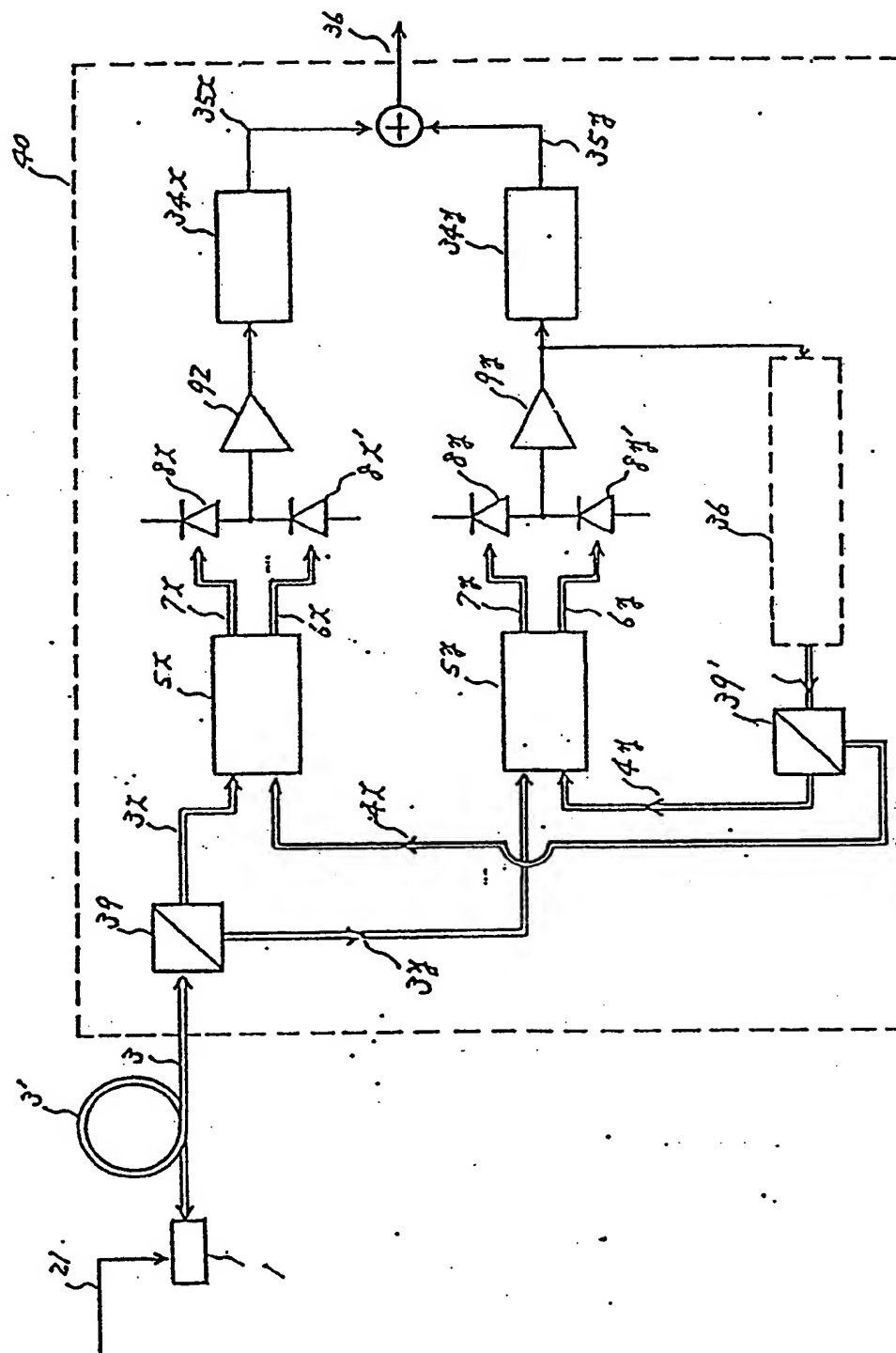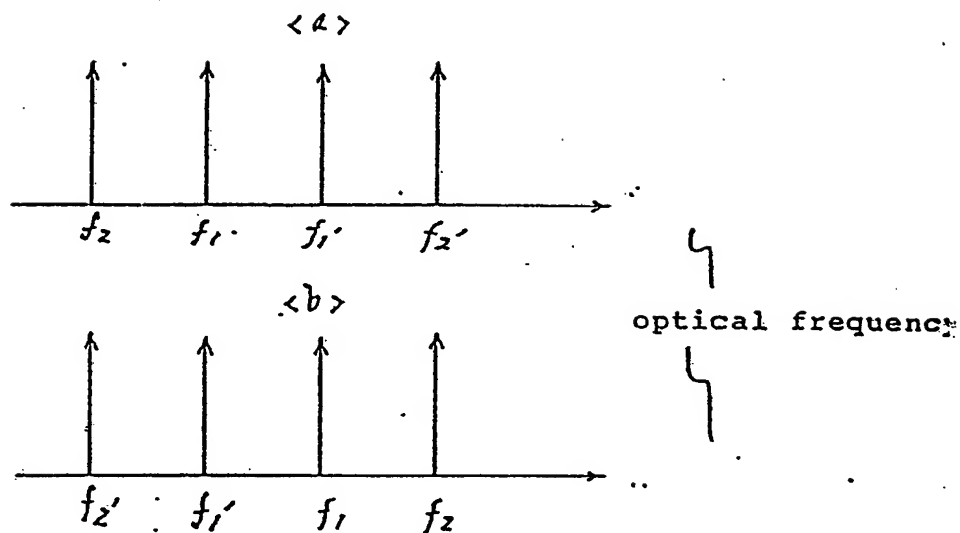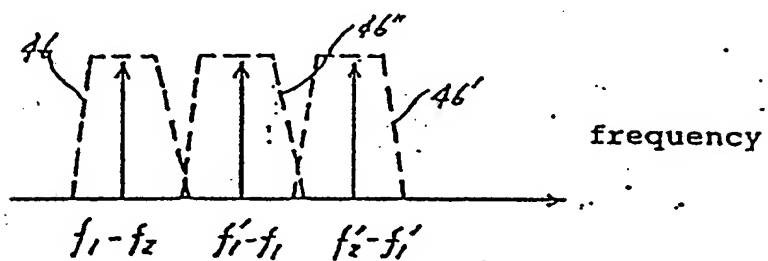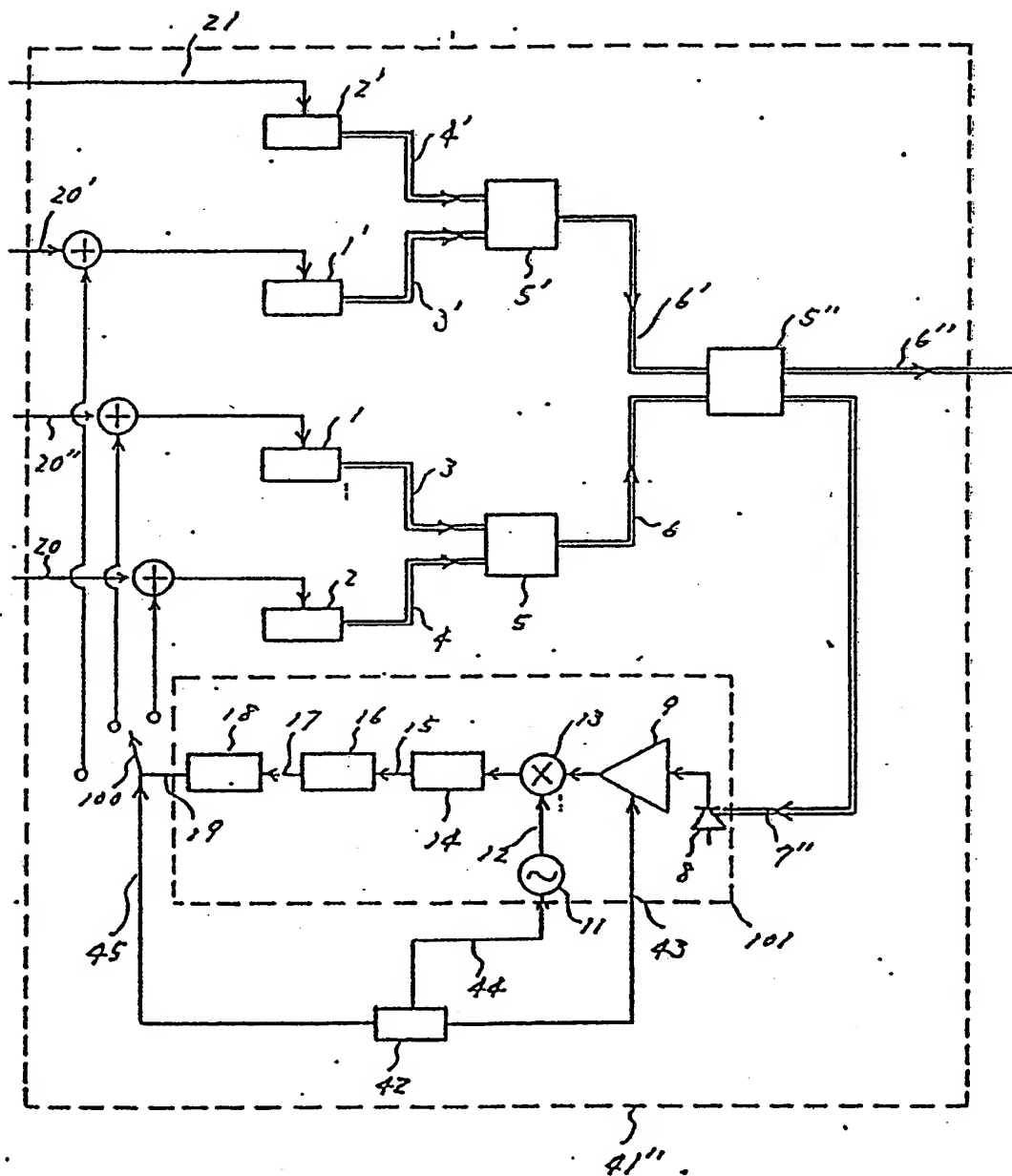
Fig. 16

Fig. 17
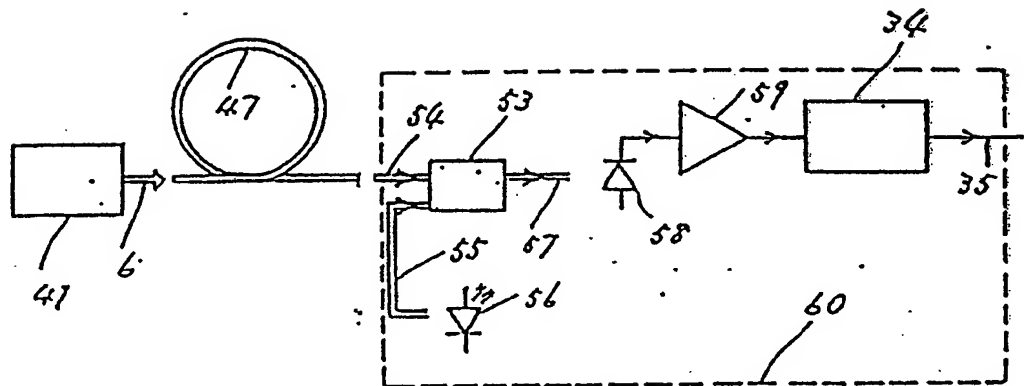
Fig. 18

Fig. 19

Fig. 21
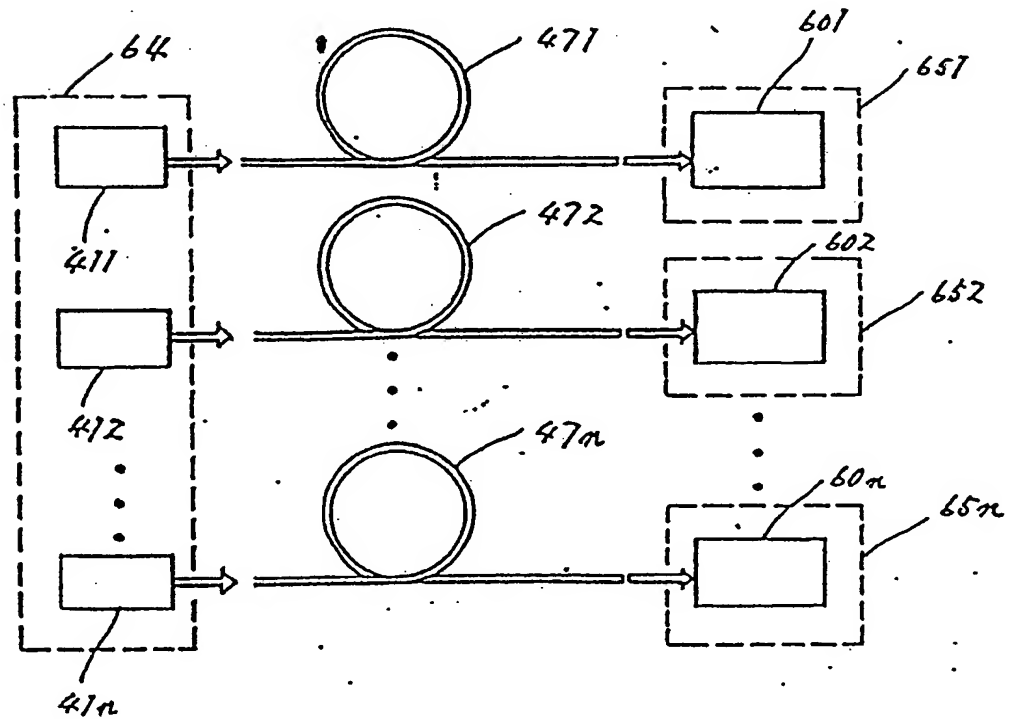
Fig. 20
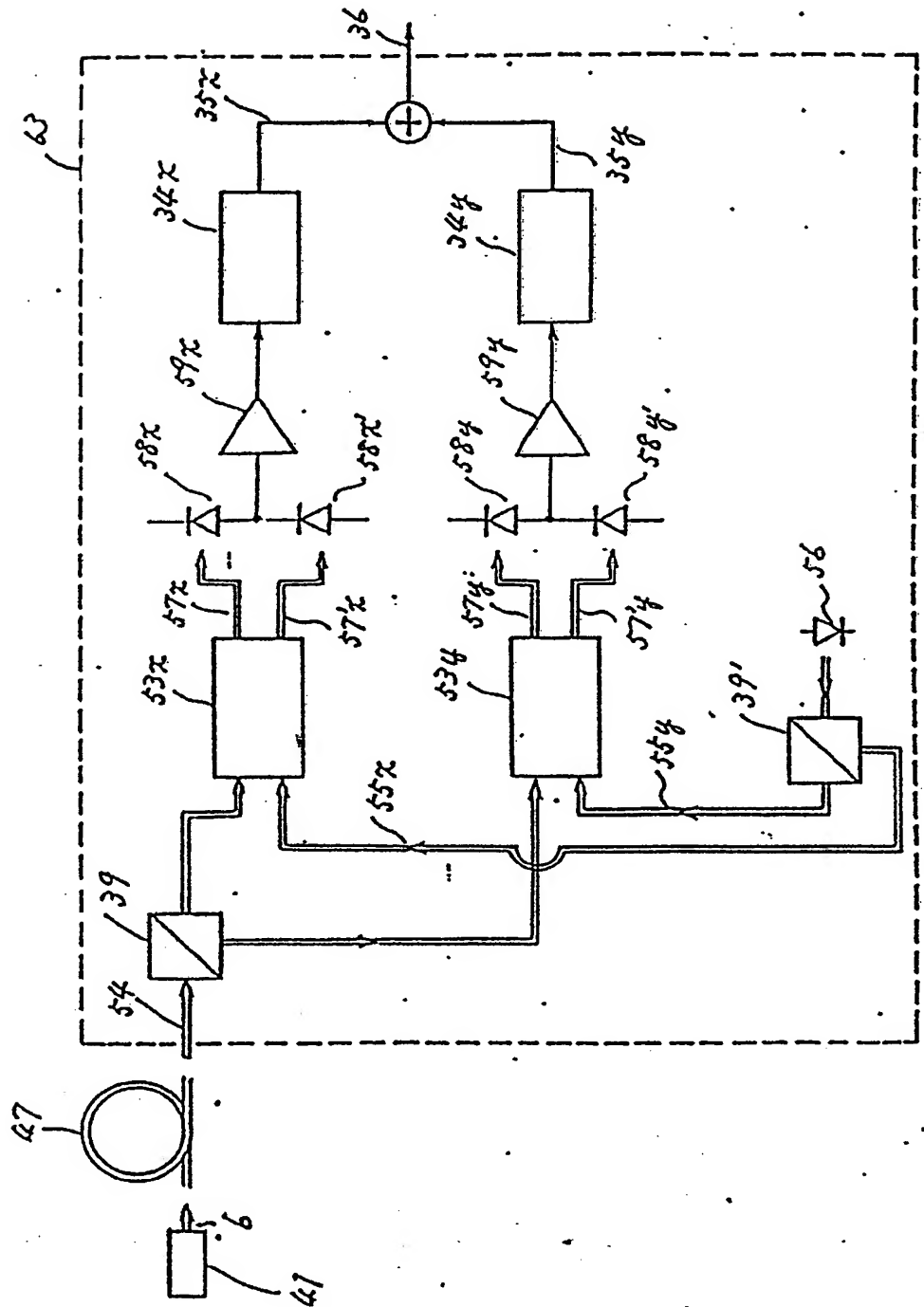
EUROPEAN PATENT APPLICATION

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(54) **Frequency separation stabilization method for optical heterodyne or optical homodyne communication.**

(57) A system for stabilizing a frequency separation between two coherent light sources includes first and second current sources driving first and second semiconductor laser devices (1, 2) with first and second frequencies (f1, f2), respectively. Light from the semiconductor laser devices (1, 2) is combined in a multiplexer (5), and a portion (7) of the combined light is diverted to an optical-to-electrical transducer (8). The output current (10) from the transducer (8) is combined with an output (12) of a local oscillator (11) driven at a selected frequency. A combined signal is then filtered by a filter (14) and used to form a control signal of one of the semiconductor laser devices (2).

Fig. 1

DOCUMENTS CONSIDERED TO BE RELEVANT

Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	CLASSIFICATION OF THE APPLICATION (Int. Cl.5)
Y	EUROPEAN CONFERENCE ON OPTICAL COMMUNICATION 21 September 1982, CANNES, FR pages 174 - 178; STEELE ET. AL.: 'Electronic controll of a difference frequency between two semiconductor lasers' — — —	1,2,6	H 04 B 10/14
A	* the whole document *	10,11,12, 17,18	
	— — —		
Y	APPLIED PHYSICS B vol. B 35, no. 7, November 1984, HEIDELBERG, DE pages 119 - 122; WHITFORD: 'Simultaneous phase-lock of five CO2 lasers to a primary Cs frequency standard' * figure 2 * * — — —	1,2,6	
A	MO-A- (Topics in applied physics vol. 59)' 1987 , SPRINGER VERLAG , BERLIN, DE * page 13, paragraph B * * — — —	3,4,15	
A	JOURNAL OF OPTICAL COMMUNICATION vol. 2, no. 3, September 1981, BERLIN pages 82 - 88; OKOSHI, KIMURA: 'Heterodyne-type optical communication' * page 83, paragraph 4 * * * page 86, paragraph 7 * * — — — — —	5,8,9,16	
			TECHNICAL FIELDS SEARCHED (Int. Cl.5)
			H 04 B H 01 S H 04 J

The present search report has been drawn up for all claims

Place of search	Date of completion of search	Examiner
The Hague	28 October 91	WAGNER U.

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